

# The Myths, Realities and Futures of NOMA

## A Historic Perspective on FDMA, TDMA, CDMA, MC-CDMA, SDMA, IDMA, CCDMA and All That...

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# Acknowledgements

- Sincere thanks for the cordial invitation to this vivacious journey to 5G Square...
- To the Team back at 'base' in Southampton, UK;
- To the Sponsors: EPSRC, the ERC, RS;
- To the entire VTS/NOMA Community;

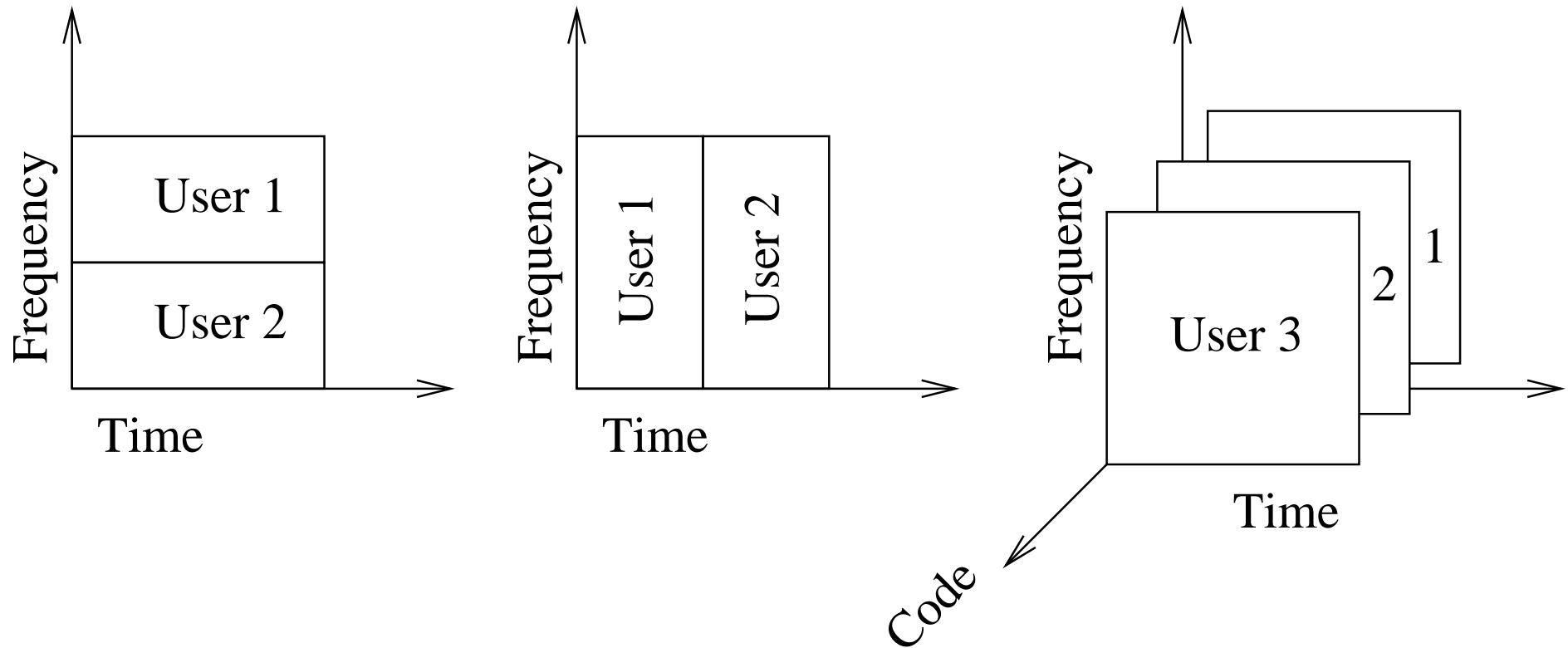
# Evolution from Marconi to NOMA & 5G...



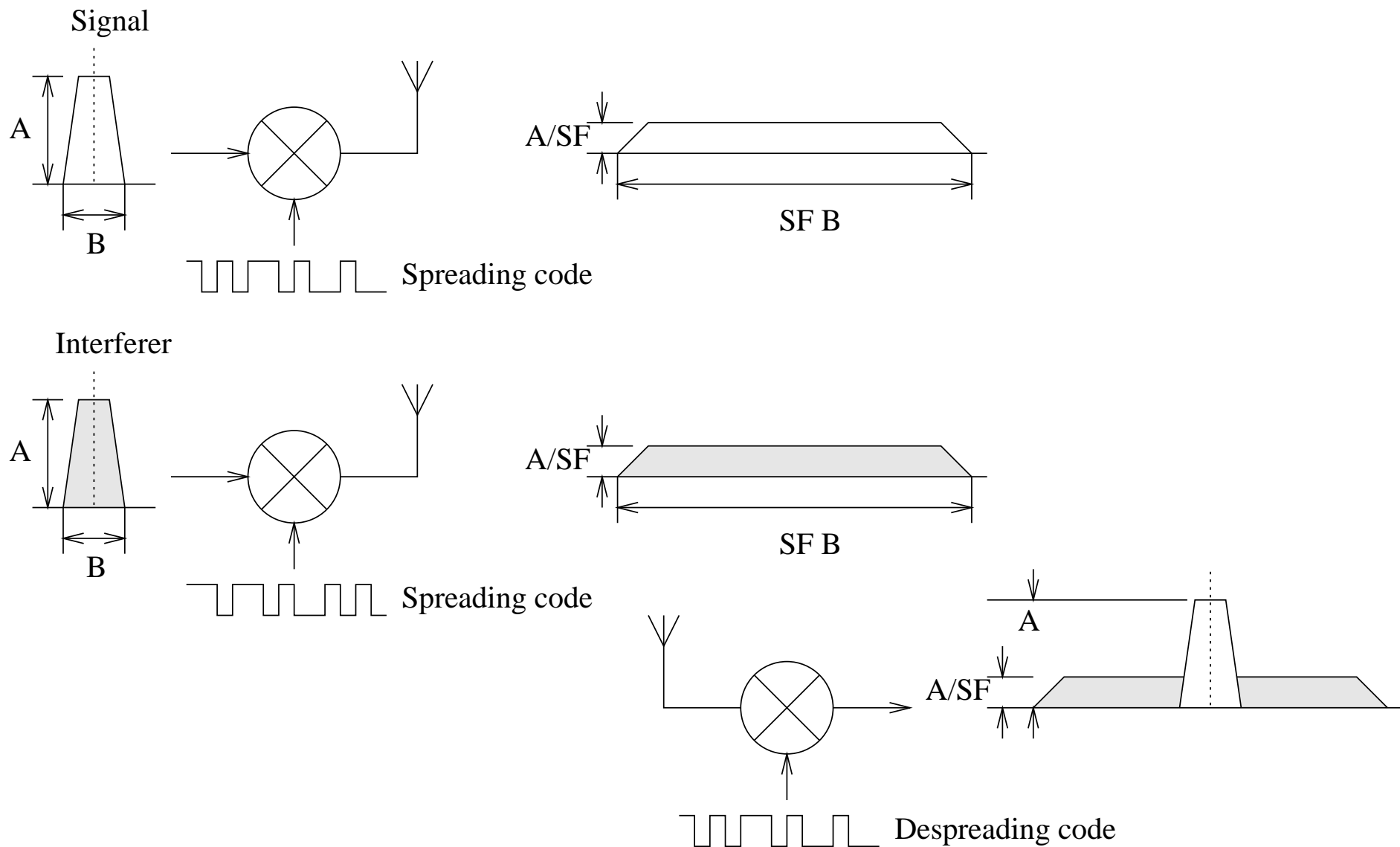
# The Myths, The Realities & The Challenges

- **The Myth:** Channel capacity is arbitrarily approached - **zero error and cost-effective, flawless 'telepresence' for anyone, anywhere, anytime!**
- **The Reality:** The moment we leave the office, our ability to access multimedia services becomes desperately limited - if not unfeasible - especially on the move!
- The **channel quality fluctuates by as much as 40dB**, hence it is unrealistic to expect that any fixed-mode wireless system provides a near-constant QoS - especially in the face of realistic channel estimation and synchronization...
- **The Challenges...**

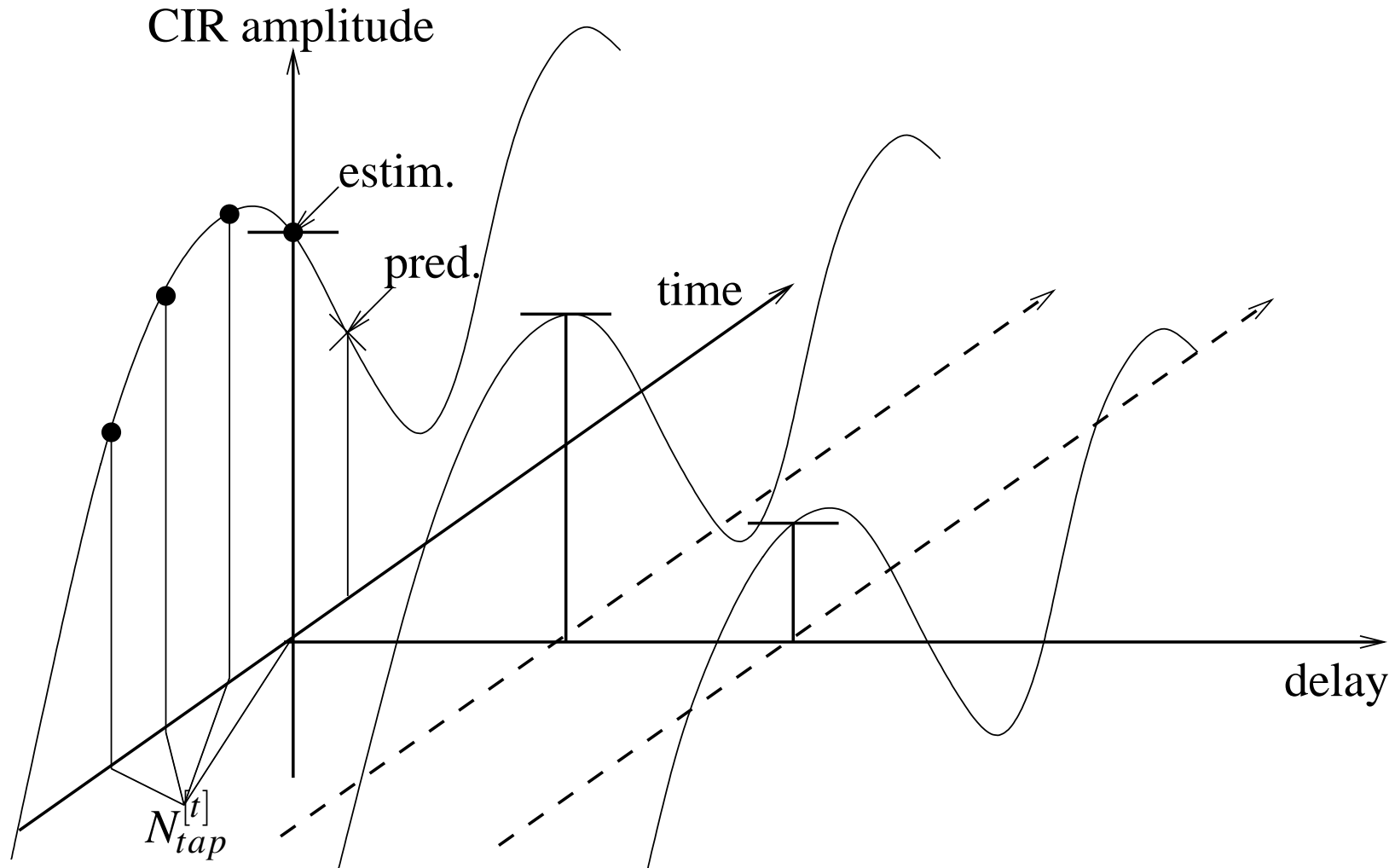
# Orthogonal multiple access schemes: FDMA, TDMA and CDMA



# Intentional DS-SS Spreading

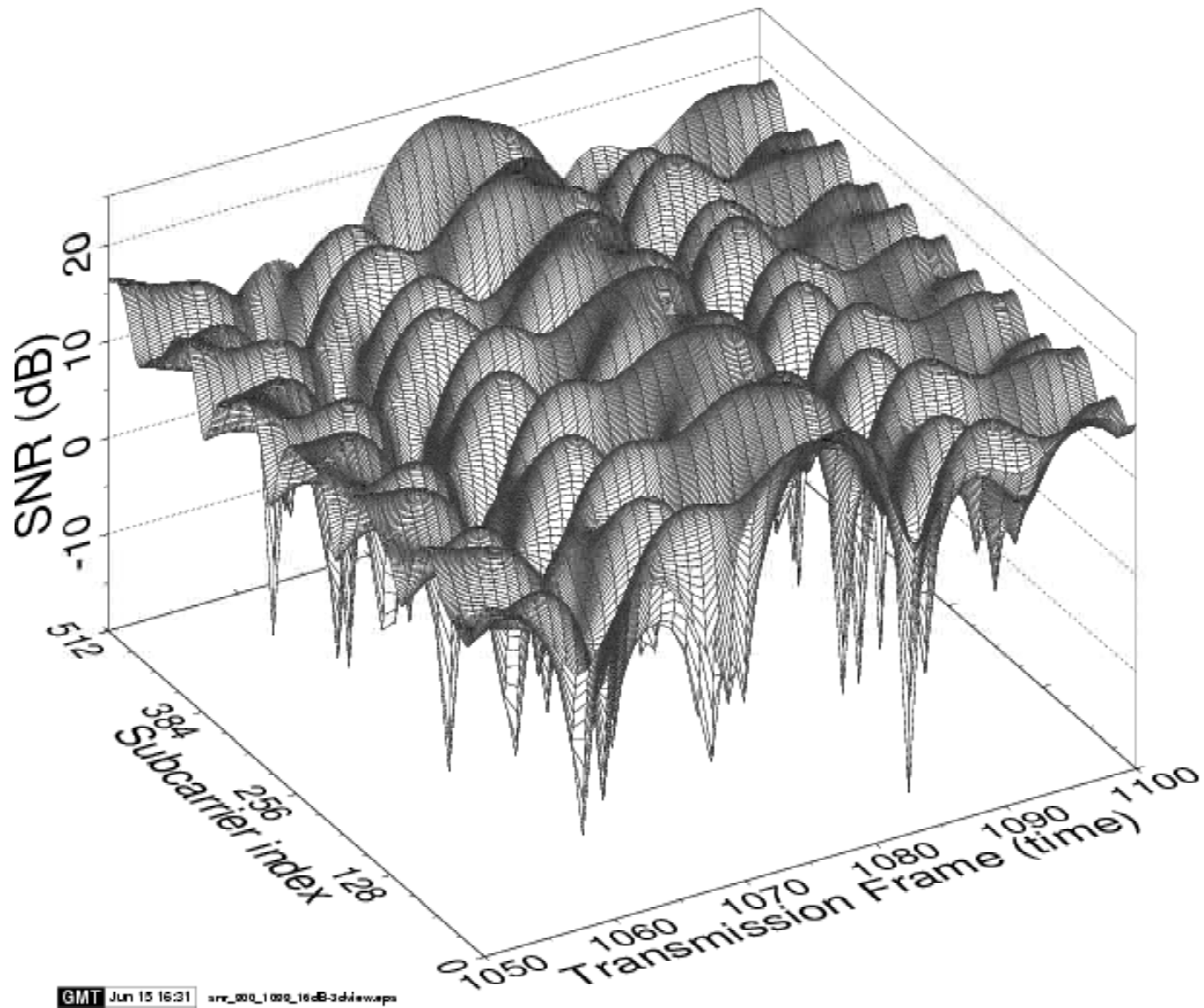


# Unintentional Channel-Induced Spreading

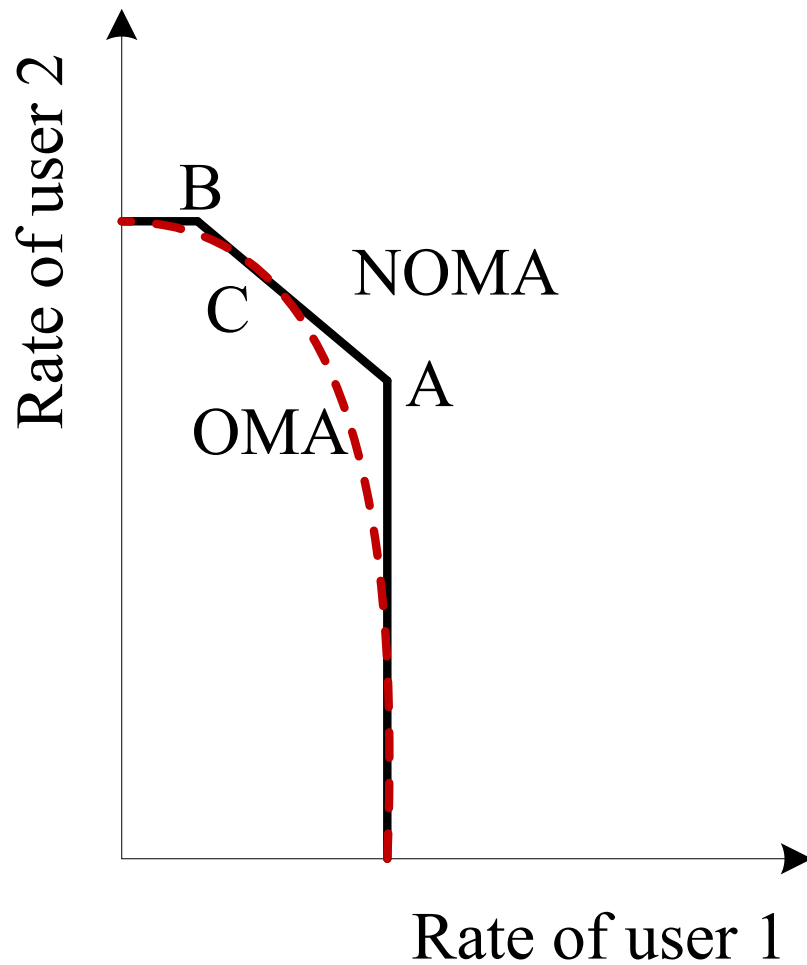




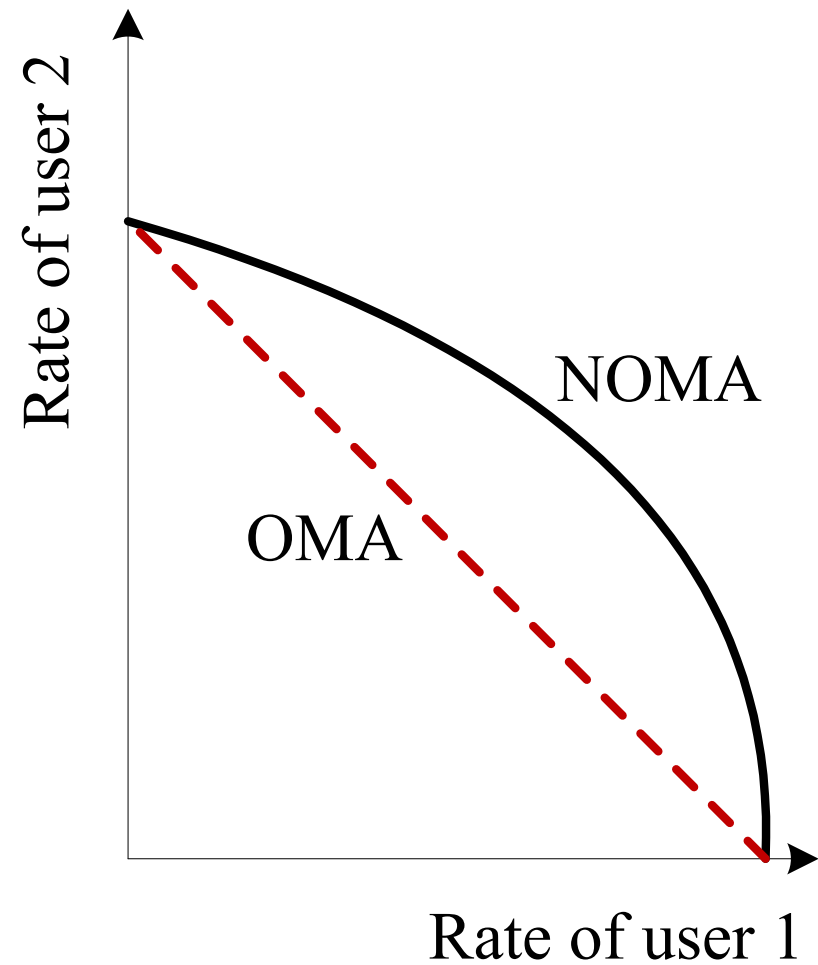
# Unintentional Spreading in the FD



# Capacity of OMA vs. NOMA in AWGN channel: (a) Uplink; (b) Downlink.



(a)



(b)

# Diverse NOMA contributions from Southampton Wireless

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- S. Chen, L. Hanzo, and A. Livingstone, “MBER space-time decision feedback equalization assisted multiuser detection for multiple antenna aided SDMA systems,” *IEEE Trans. Signal Process.*, vol. 54, no. 8, pp. 3090–3098, Aug. 2006.

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- A. Wolfgang, J. Akhtman, S. Chen, and L. Hanzo, “Reduced-complexity near-maximum-likelihood detection for decision feedback assisted space-time equalization,” *IEEE Trans. Wireless Commun.*, vol. 6, no. 7, pp. 2407–2411, Jul. 2007.
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- S. Chen, A. Livingstone, H. Q. Du, and L. Hanzo, “Adaptive minimum symbol error rate beamforming assisted detection for quadrature amplitude modulation,” *IEEE Trans. Wireless Commun.*, vol. 7, no. 4, pp. 1140–1145, Apr. 2008.
- J. Zhang, S. Chen, X. Mu, and L. Hanzo, “Turbo multi-user detection for OFDM/SDMA systems relying on differential evolution aided iterative channel estimation,” *IEEE*

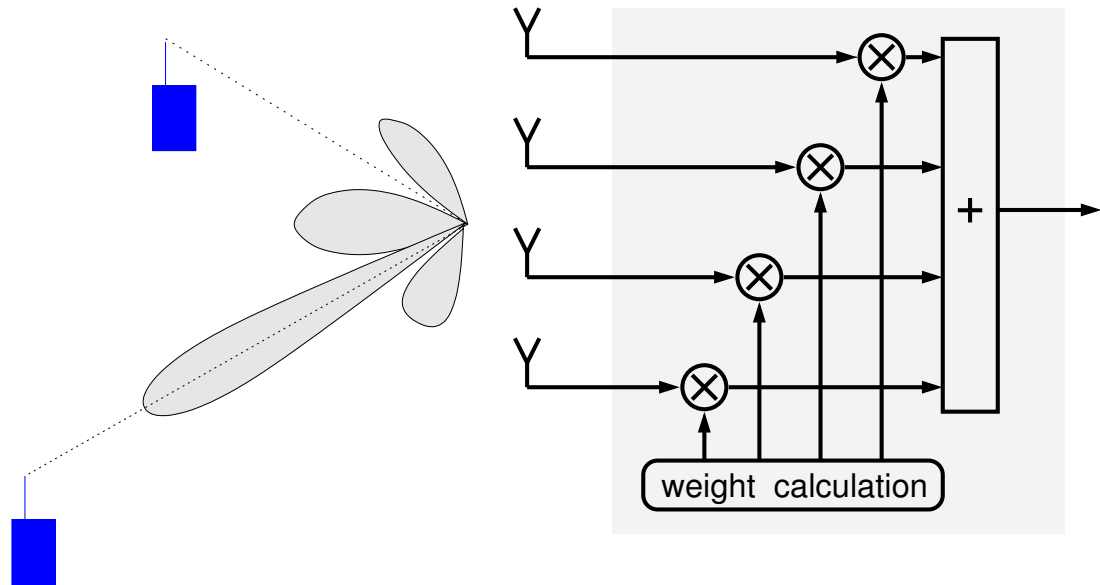
*Trans. Commun.*, vol. 60, no. 6, pp. 1621–1633, Jun. 2012.

- J. Zhang, S. Chen, X. Mu, and L. Hanzo, “Joint channel estimation and multi-user detection for SDMA/OFDM based on dual repeated weighted boosting search,” *IEEE Trans. Veh. Technol.*, vol. 60, no. 7, pp. 3265–3275, Jun. 2011.
- C.-Y. Wei, J. Akhtman, S.-X. Ng, and L. Hanzo, “Iterative near-maximum-likelihood detection in rank-deficient downlink SDMA systems,” *IEEE Trans. Veh. Technol.*, vol. 57, no. 1, pp. 653–657, Jan. 2008.
- A. Wolfgang, J. Akhtman, S. Chen, and L. Hanzo, “Iterative MIMO detection for rank-deficient systems,” *IEEE Signal Process. Lett.*, vol. 13, no. 11, pp. 699–702, Nov. 2006.
- L. Xu, S. Chen, and L. Hanzo, “EXIT chart analysis aided turbo MUD designs for the rank-deficient multiple antenna assisted OFDM uplink,” *IEEE Trans. Wireless Commun.*, vol. 7, no. 6, pp. 2039–2044, Jun. 2008.

# ***NOMA Beamforming Example***

# Uplink/Downlink Beamforming

- Why?  
Increase of capacity
- How?  
Spatially separated interfering signals are suppressed



$$y = \mathbf{w}^H \mathbf{X}$$

# MMSE Based Beamforming

- Weights are calculated in order to minimize:

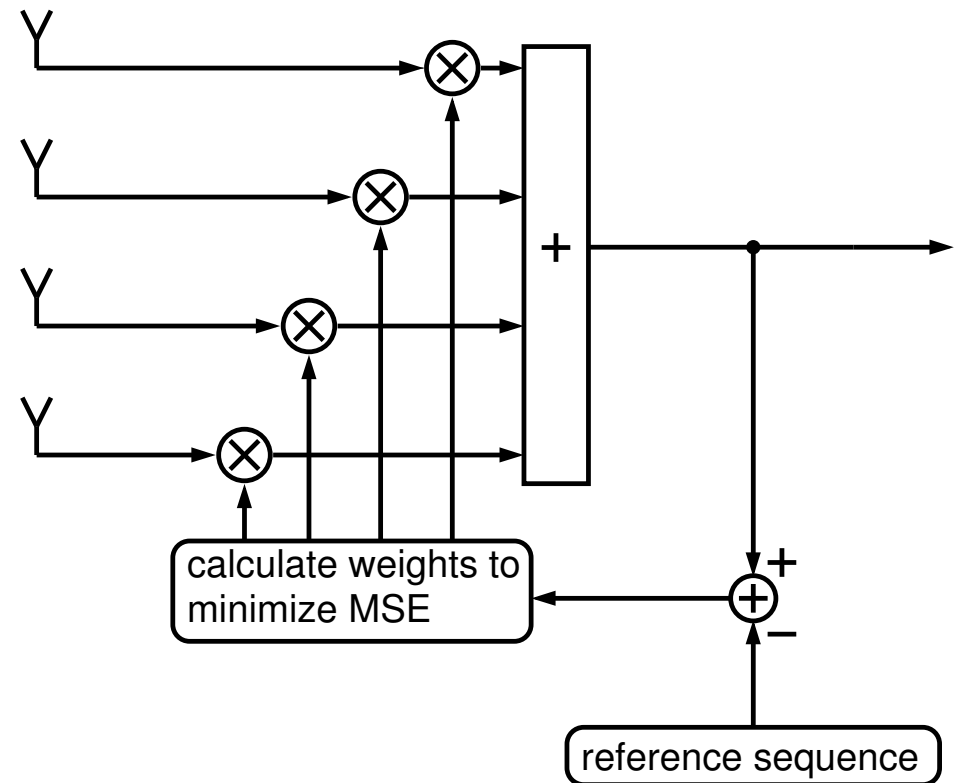
$$\varepsilon(t)^2 = (\mathbf{w}^H \mathbf{x}(t) - r(t))^2$$

$\mathbf{w}$ : Beamformer weights

$\mathbf{x}(t)$ : Channel output

$r(t)$ : Reference symbol

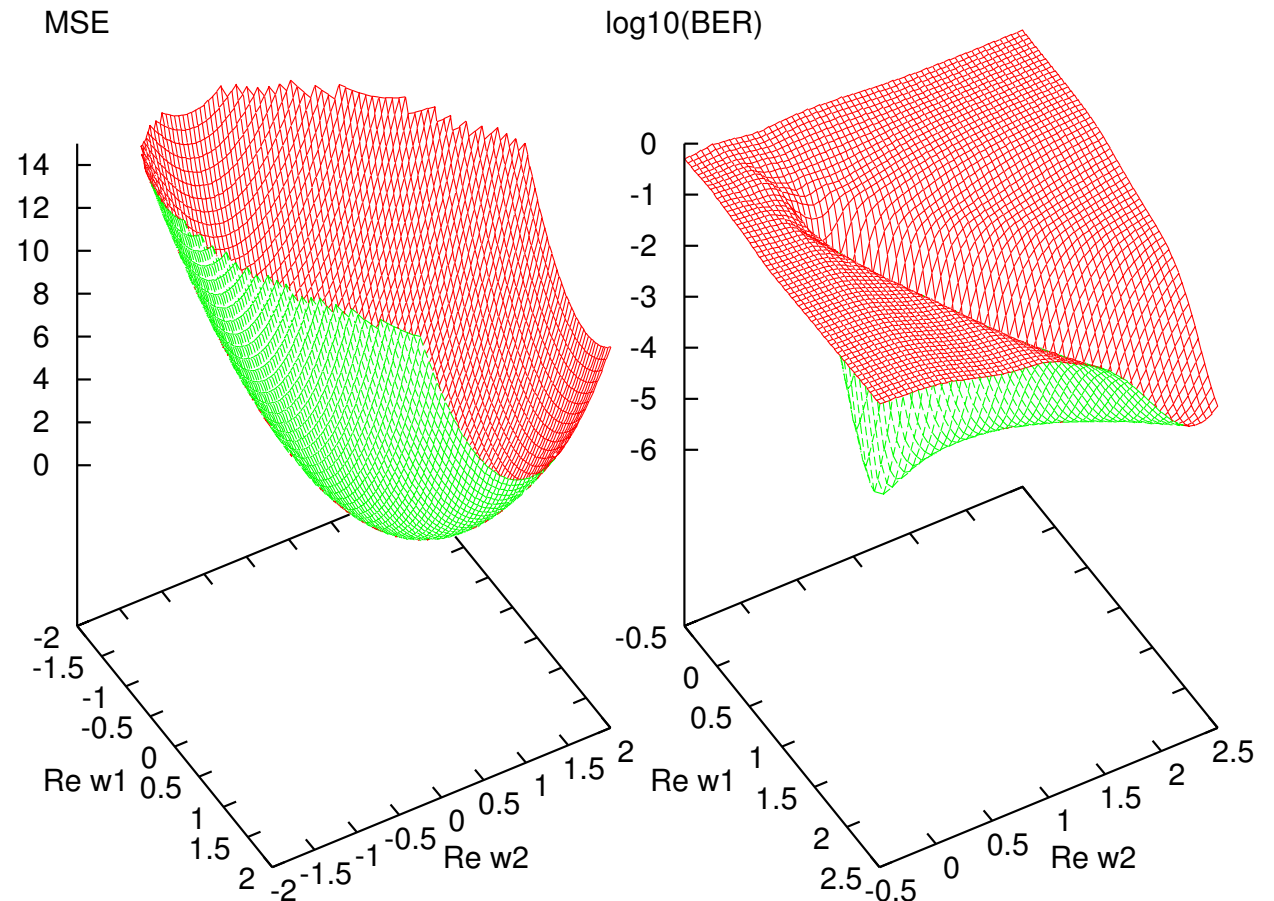
- For AWGN channels MMSE weights can be calculated using a closed form expression
- Realizations: LMS, RLS, SMI





# MSE and BER Surfaces at the Output of a [5 x 2] NOMA Beamformer

Error surfaces at the receiver's output calculated for five BPSK modulated sources having equal received power and communicating over AWGN channels at SNR=10 dB.

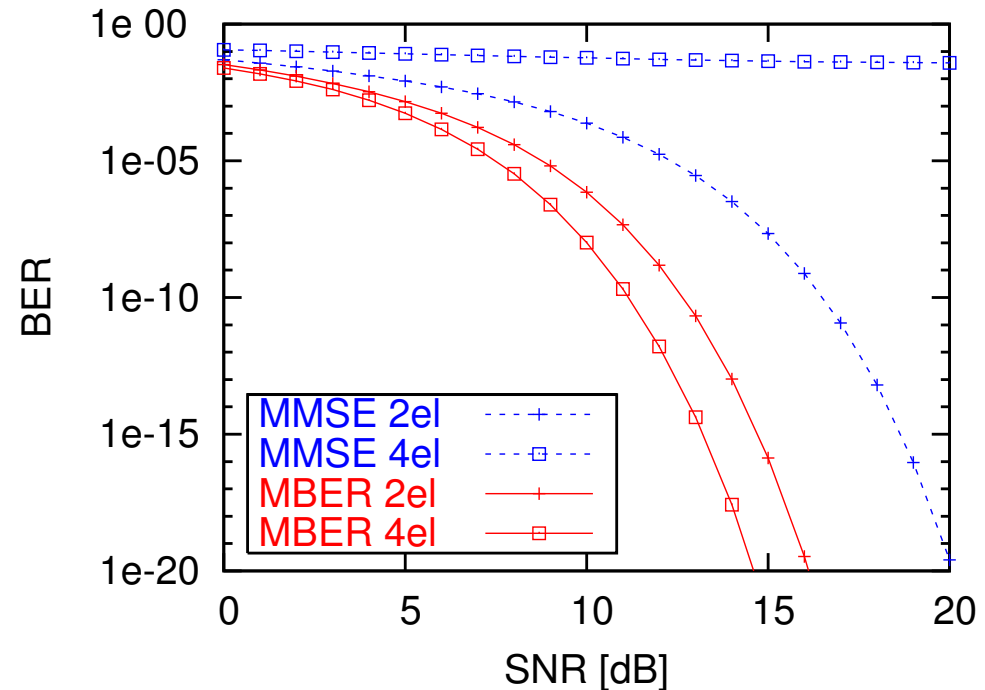


The imaginary part of both weights of the 2-element array was fixed.

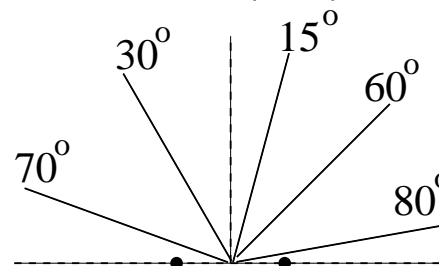
- **Is MBER detection the next stage of evolution?**

# MMSE vs MBER NOMA Beamforming

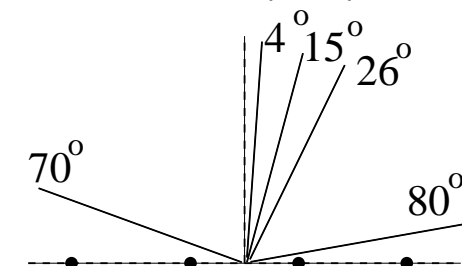
- Test case: BPSK modulated sources having equal received power and communicating over AWGN channels
- MMSE solution calculated analytically
- MBER solution obtained with the aid of conjugate gradient algorithm



Scenario S (2el.)

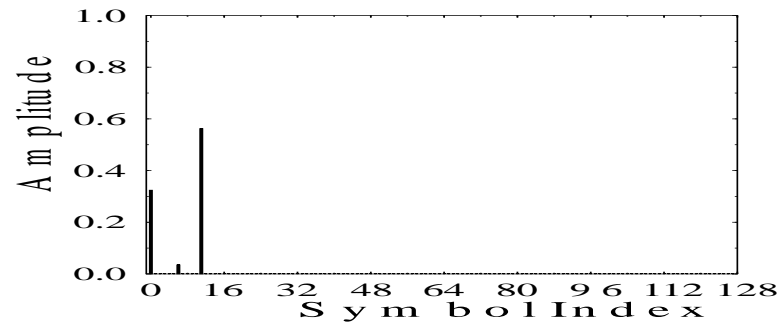


Scenario U (4el.)

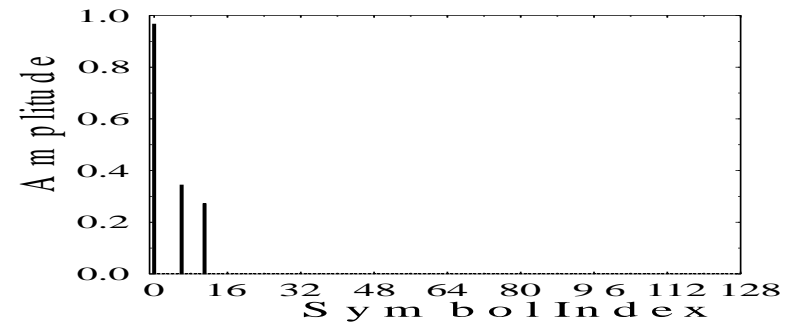


# ***NOMA SDMA Example***

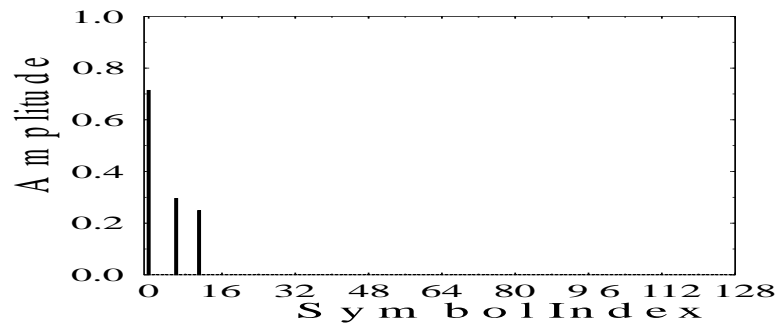
# Evolution from CDMA-NOMA to SDMA-NOMA



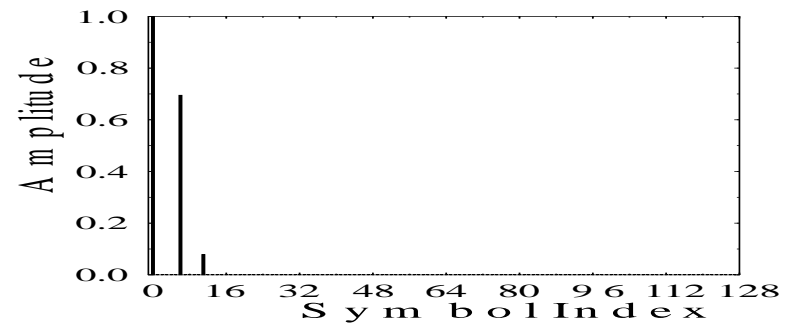
(a) CIR 1: user 1, antenna 1



(b) CIR 2: user 1, antenna 2

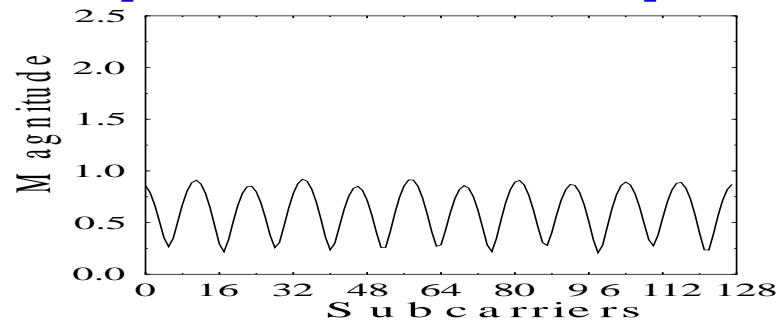


(c) CIR 3: user 2, antenna 1

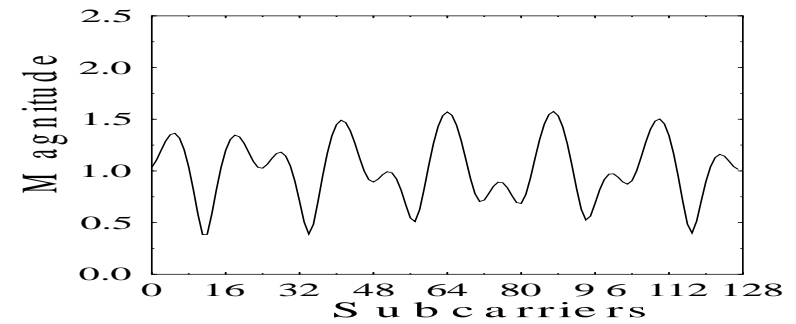


(d) CIR 4: user 2, antenna 2

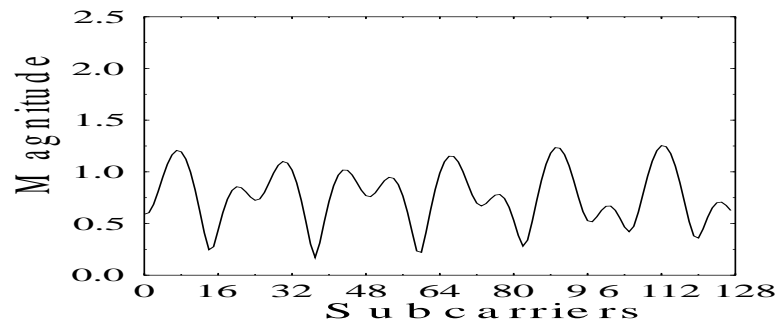
# Unique Stream-Specific FD-CTF of CIRs 1-4



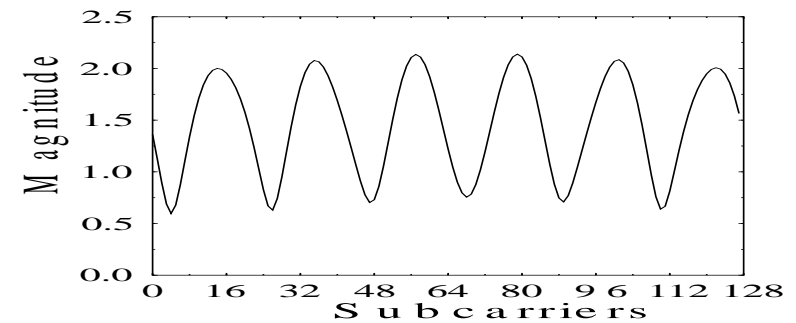
(e) CTF 1: user 1, antenna 1



(f) CTF 2: user 1, antenna 2



(g) CTF 3: user 2, antenna 1



(h) CTF 4: user 2, antenna 2

Figure 1: FD Channel transfer functions (FD-CTF) for the CIRs seen in Figure 20 (a) FD-CTF 1, (b) FD-CTF 2, (c) FD-CTF 3, and (d) FD-CTF 4.

# Quantum-Search Aided MUD in NOMA

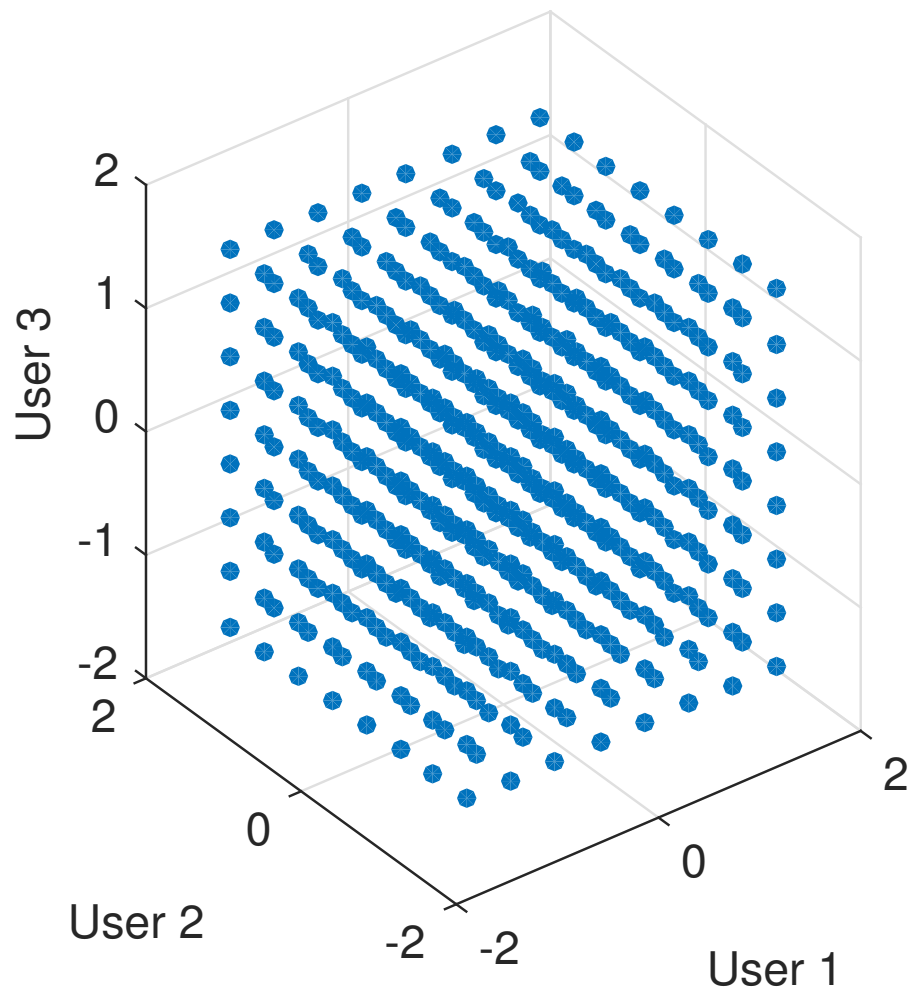
Multiple Access	SDMA-OFDM
Number of Users	$U = 3$
Number of AEs at the BS	$P = 1$
<b>Normalized User-Load</b>	$U_L = U_q/P = 3$
Modulation	8-PAM $M = 8$
$E_b/N_0$	0 dB
Channel Code	Turbo Convolutional Code, 8 trellis states, $R = 1/2$
Channel Model	Extended Typical Urban (ETU)
Mobile Velocity	$v = 130$ km/h
Carrier Frequency	$f_c = 2.5$ GHz
Sampling Frequency	$f_s = 15.36$ GHz (77 delay taps)
Doppler Frequency	$f_d = 70$ Hz
Number of Subcarriers	$Q = 1024$
Cyclic Prefix	CP = 128
Interleaver Length	10240 bits per user
Channel Estimation	Perfect

# Quantum-Search Aided MUD in NOMA

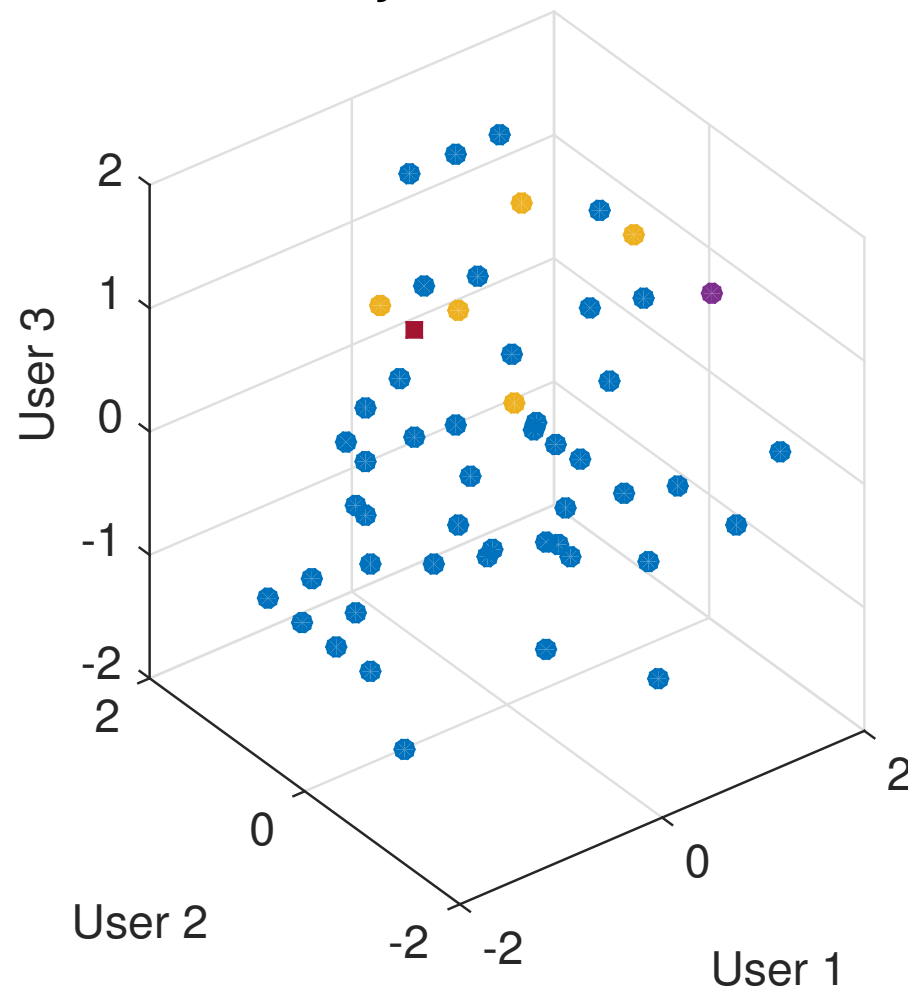
- There are  $8^3 = 512$  symbols in the full constellation, while 53 and 46 symbols are obtained by the randomly-initialized and ZF-initialized DHA, respectively.
- The purple circle denotes the random initial input, or the ZF detector's output, which may be used as an initial input. The ZF is as bad as the random one in this rank-deficient scenario.
- By using the DHA, we find symbols better than the previously found symbols, which are denoted by the yellow circles in the 3D figure.
- But we also find symbols that are "worse" than the previously found symbols, as represented by the blue circles in the 3D figure.
- The red square is the optimal symbol which is eventually found.

# Dürr-Høyer MUD for CDMA/SDMA NOMA - Userload=2

Full Constellation



Randomly Initialized DHA





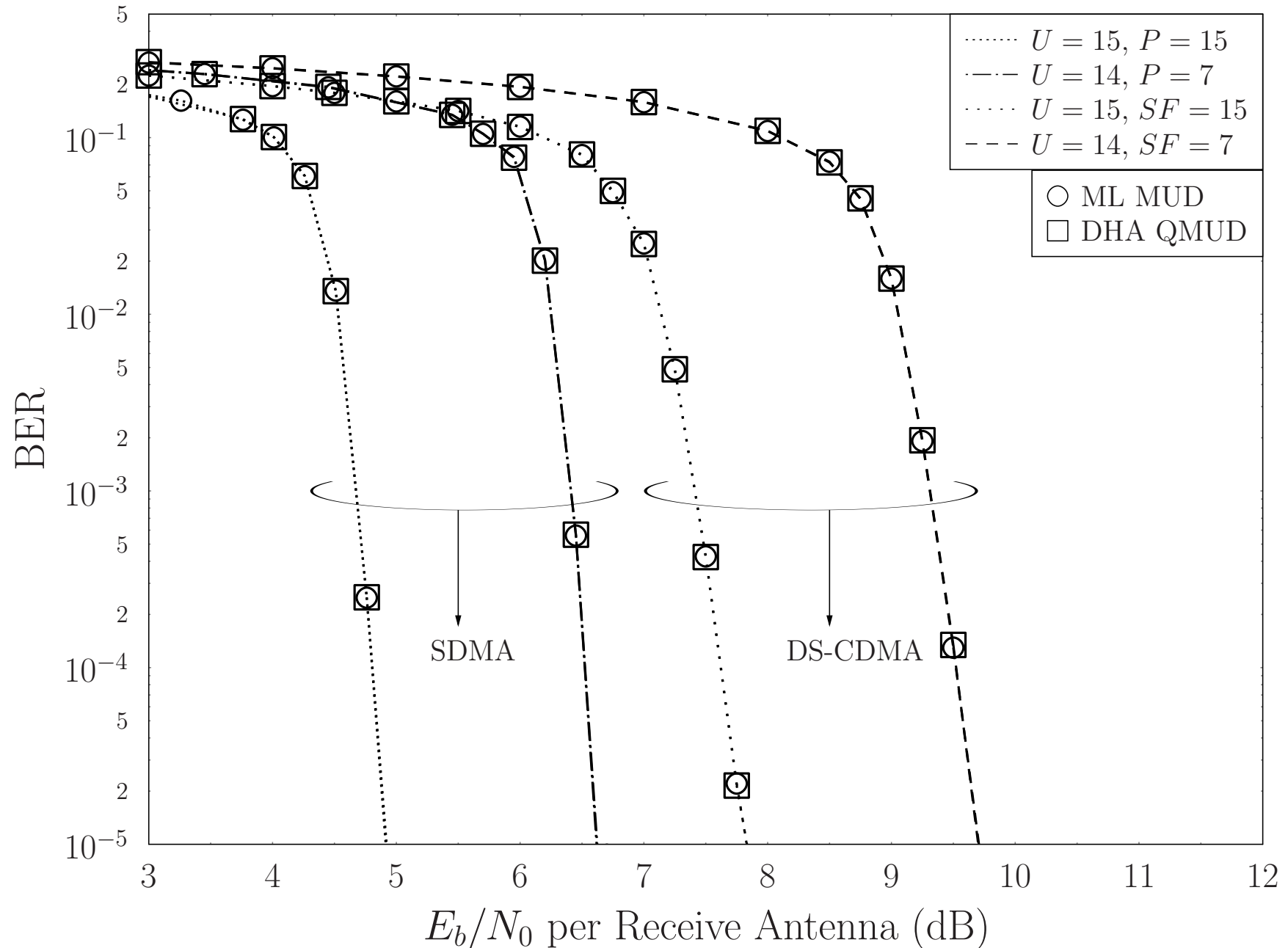
# Quantum Computing Meets MUD

## *NOMA CDMA vs SDMA*

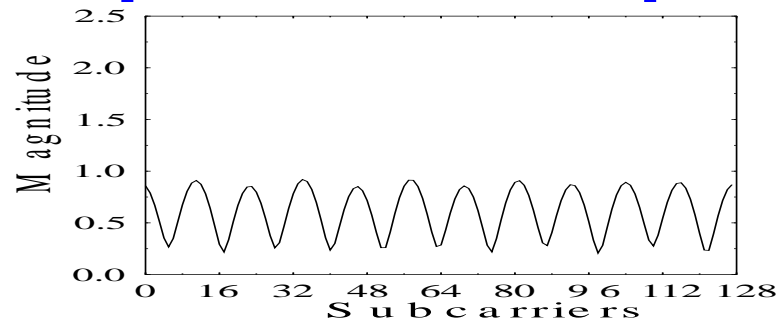
# DS-CDMA vs SDMA NOMA Systems

	System 1	System 2	System 3	System 4
Number of Users	$U = 14$	$U = 14$	$U = 15$	$U = 15$
Multiple Access Scheme	DS-CDMA	SDMA	DS-CDMA	SDMA
Number of AEs at the BS	$P = 1$	$P = 7$	$P = 1$	$P = 15$
Spreading Factor	$SF = 7$	N/A	$SF = 15$	N/A
Spreading Codes	m-sequences	N/A	Gold Codes	N/A
<b>Normalized User Load</b>	$U_L = 2$	$U_L = 2$	$U_L = 1$	$U_L = 1$
Bit-based Interleaver Length	42 000	42 000	40 000	40 000
Number of AEs per User		$N_{Tx} = 1$		
Modulation		BPSK $M = 2$		
Channel Code		Turbo Code, $R = 1/2$ , 8 Trellis states		
		$I_{inner} = 4$ iterations		
Channel		Uncorrelated Rayleigh Channel		
Channel Estimation		Perfect		

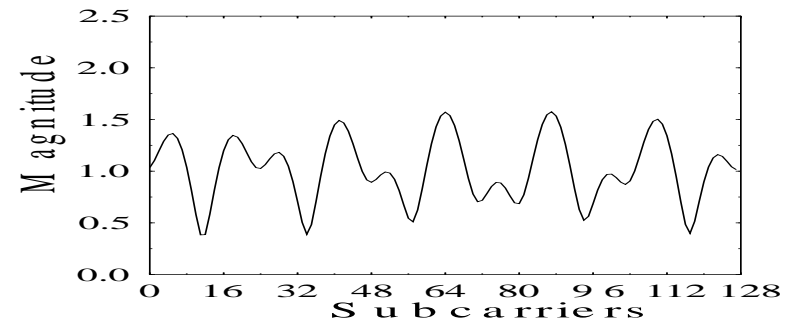
# Dürr-Høyer CDMA/SDMA NOMA AT Userload=2



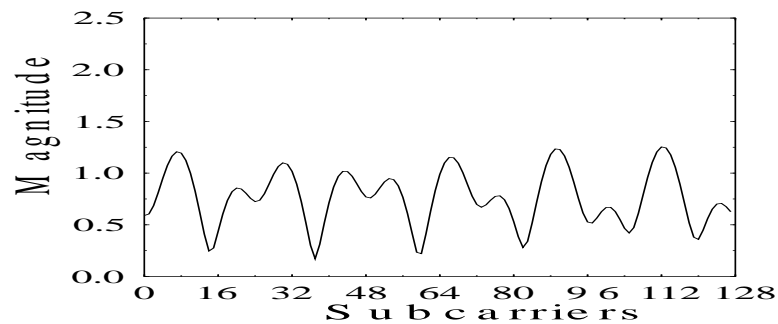
# Unique Stream-Specific FD-CTF of CIRs 1-4



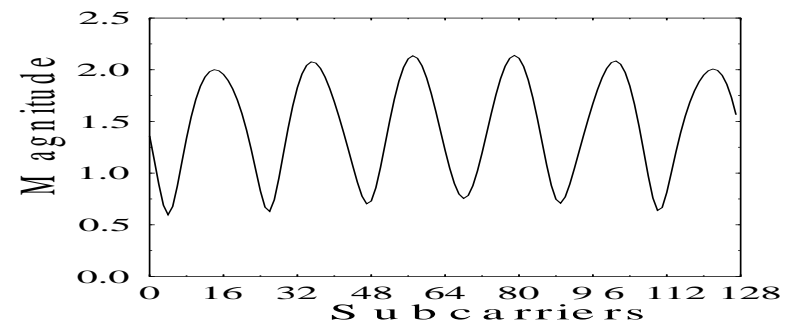
(a) CTF 1: user 1, antenna 1



(b) CTF 2: user 1, antenna 2



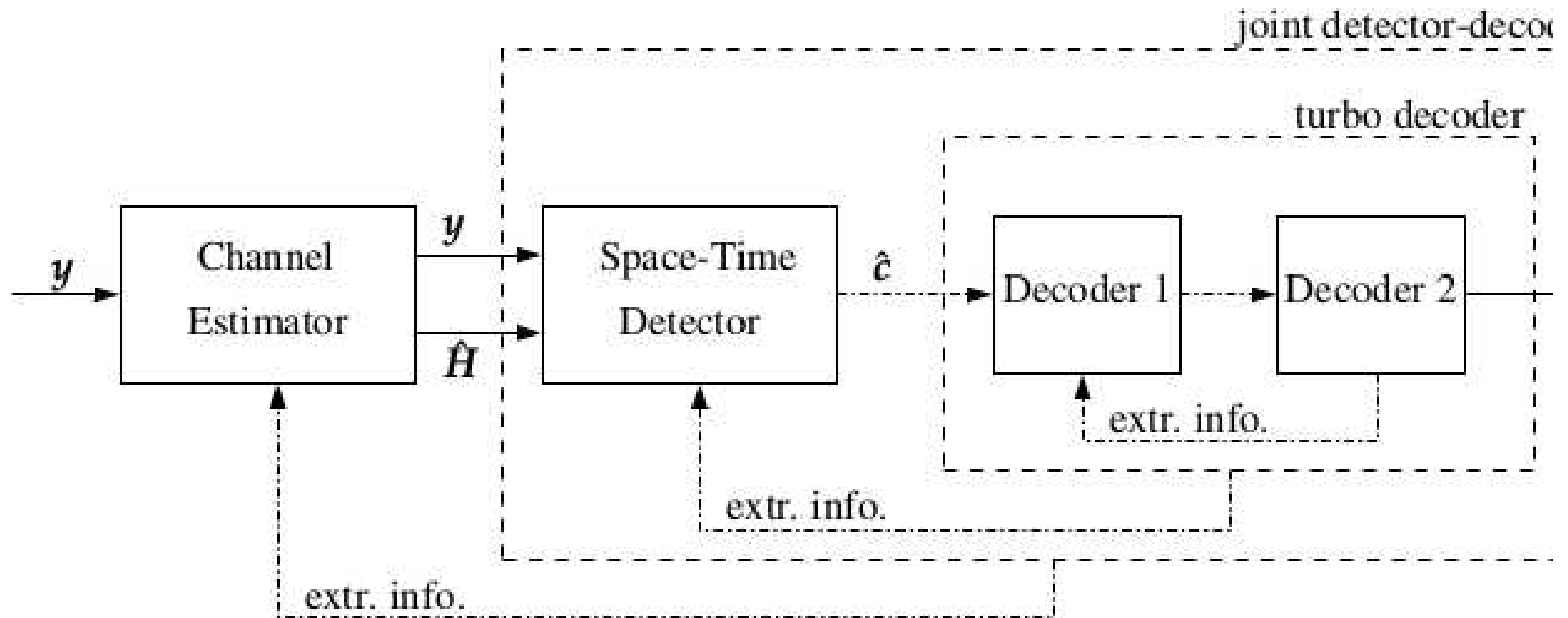
(c) CTF 3: user 2, antenna 1



(d) CTF 4: user 2, antenna 2

Figure 2: FD Channel transfer functions (FD-CTF) for the CIRs seen in Figure 20 (a) FD-CTF 1, (b) FD-CTF 2, (c) FD-CTF 3, and (d) FD-CTF 4.

# Iterative Joint Channel & Data Estimation Turbo-Receivers for NOMA



***Thank you...***